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1 Introduction

Recent years have seen major advances in photon based quantum technologies¹ such as deep space optical communication (DSOC),^{2,3} quantum key distribution (QKD),^{4,5} quantum computation⁶ or state teleportation.⁷ For such applications, components such as single-photon emitters (e.g. quantum dots,⁸ NV centres in diamond⁹ or 2D materials¹⁰), spin-photon interfaces^{11,12} or single-photon detectors are needed. In terms of detectors, superconducting nanowire single-photon detectors $(SNSPDs)^{13-16}$ have prevailed over other options such as single-photon avalanche photodiodes (SPADs)¹⁷ or transition edge sensors (TES).¹⁸ Besides near-unity quantum efficiency,¹⁹ one of their outstanding performance features is the timing jitter describing the temporal resolution.²⁰ It was shown to be as low as 2.6 ps for a short (5 µm long) and thin (80 nm wide) superconducting nanowire²¹ at λ = 775 nm. For a meandering nanowire covering an area of $10 \times 10 \ \mu\text{m}^2$, values of 28 ps have been reported.²² Moreover, the timing jitter is a crucial parameter for various applications such as the characterization of single-photon emitters,²³ single-photon QKD²⁴ or pulse-position modulation,²⁵ where information is encoded in so-called time bins. The smaller the timing jitter the more time bins fit within a certain amount of time. Therefore, it is of utmost interest to advance the understanding of the various influ-

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The dependence of timing jitter of superconducting nanowire single-photon detectors on the multi-layer sample design and slew rate

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We investigated the timing jitter of superconducting nanowire single-photon detectors (SNSPDs) and found a strong dependence on the detector response. By varying the multi-layer structure, we observed changes in pulse shape which are attributed to capacitive behaviour affecting the pulse heights, rise times and consequently timing jitter. Moreover, we developed a technique to predict the timing jitter of a single device within certain limits by capturing only a single detector pulse, eliminating the need for detailed jitter measurement using a pulsed laser when a rough estimate of the timing jitter is sufficient.

ences on timing jitter, how to improve it, and how to speed up the characterization process. To analyze the origin of timing jitter, it is useful to understand on which parameters it depends. To this end, recent works have identified that the timing jitter depends on contributions from the amplifiers,²⁶ hotspot formation,²⁷ bias current and operation wavelength,²⁸ the pulse width of the laser, electrical and thermal²² noise as well as the detector geometry.^{22,29} Each of these components can be analyzed individually for instance by using a dual-readout scheme²² and can be assigned to either intrinsic²⁸ (hotspot formation, detector geometry) or extrinsic (amplifiers, pulse width, noise) properties. One parameter derived from the extrinsic properties is the signal-to-noise ratio (SNR), where You et al. were able to associate an improved jitter with an increased SNR.³⁰ In addition, Wu et al. were able to demonstrate a direct correlation between the maximum edge slope (hereafter referred to as slew rate) and timing jitter.²⁶ They compared different publications and found that lower jitter values were associated with a higher slew rate, which was confirmed by Korzh et al.²¹ In this work, we investigate this relation in more detail by analyzing the influence of different material layer combinations on the timing jitter and draw conclusions for optimized multi-layer structures for SNSPDs that enable high detection efficiency and low timing jitter. Afterwards we show how to calibrate a measuring system to determine the jitter from the slew rate, *i.e.* the ratio of pulse height and rise time.

2 Experimental

All measurements were carried out with a Janis cryogenic probestation at 4.5 K. The detectors are made of niobium

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titanium nitride (NbTiN), with a typical thickness of 8 nm, nanowire width of 100 nm and an areal fill factor of 33% covering an area of $10 \times 10 \ \mu m^2$. For the efficiency measurements a continuous wave (CW) laser (780 nm) was used to illuminate the detectors. In Fig. 1(a) a typical detector response is shown. It consists of a fast rising edge and an exponential decay following $\tau_{\text{fall}} = L_k/Z_{\text{load}}$,³¹ where Z_{load} is the load impedance (typically 50 Ω) and $L_k = \frac{\hbar R_{\text{device}}}{\pi 1.76 k_B T_c}$,³² the kinetic inductance, which was 194 nH for the detectors throughout this study. Here R_{device} describes the device resistance and T_{c} corresponds to the switching temperature of the superconductor. For the timing measurements a pulsed laser (pulse width <3 ps, at 850 nm) was used. The laser signal is divided between a ultrafast photodiode (UPD-15-IR2-FC, start signal) with a total rise time of less than 15 ps and the detector unit consisting of the detector and two room temperature amplifiers (combined 53.5 dB, stop signal). Both channels are connected to a sampling scope (MSO64) to perform a delay measurement. In Fig. 1(b), the normalized occurrences of the delay measurement are plotted as a function of time. The corresponding full width at half maximum (FWHM) of the obtained Gaussian distribution represents the devices timing jitter (here: 28.6 ps).

To characterise the timing jitter of the setup, an arbitrary waveform generator (AWG) was used that mimics the detector signal before it is attenuated and re-amplified. Apart from the signal source, the measurement setup matches that of the detector measurement. Depending on the amplified pulse height we observe a drop from more than 60 ps at 0.1 V over 16.4 ps at 0.4 V to just under 10 ps at 0.7 V (cf. Table 1). Hence, the timing jitter of the setup (cf. Table 1), mainly originating from the room temperature amplifiers, is significantly smaller than the measured device jitter values (cp. Fig. 1) ensuring that it did not influence the presented results. This can be analyzed using $t_{\text{tot}} = \sqrt{\sum t_i^2}$, where t_{tot} is the measured timing jitter and t_i are the different contributions. The uncertainty of the data presented in Table 1 is mainly limited by the internal device jitter of the sampling scope (1.5 ps) and the fast photodiode approximately to be less than 5 ps.

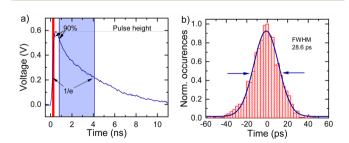


Fig. 1 Detector response upon photon absorption. (a) Typical voltage pulse showing a fast rising edge and exponential decay. The rise time is defined as the time between 1/*e* and 90% of the maximum pulse height. (b) Timing jitter determined using a pulsed laser to perform a start-stop measurement between a photodiode (start signal) and the detector unit (stop signal). The resulting normalized occurrences are plotted as a function of time. The timing jitter is determined as the FWHM of the obtained Gaussian distribution.

Table 1 Timing jitter of the measurement setup, characterized by mimicking the detector signal with an arbitrary waveform generator before it is attenuated and amplified again. For the resulting pulse height, the timing jitter is measured as depicted in Fig. 1(b) and drops from more than 60 ps at 0.1 V over 16.4 ps at 0.4 V to just under 10 ps at 0.7 V

Pulse height (V)	Jitter (ps)	Pulse height (V)	Jitter (ps)
0.1	61.1	0.4	16.4
0.15	41.9	0.45	14.4
0.2	31.6	0.5	12.6
0.25	26.4	0.55	11.6
0.3	22.0	0.6	10.8
0.35	19.0	0.7	9.6

3 Results

To investigate the dependence of the pulse shape on the timing jitter, samples with different multi-layer structures were fabricated. While most detectors were produced on a full-chip sized (FCS) gold mirror (10 nm Ti/50 nm Au) with varying silicon dioxide (SiO₂) layer thicknesses to form a single-sided cavity, detectors were furthermore fabricated on small (100 μ m diameter) gold mirrors (hereafter: Au*) with 105 nm SiO₂ and on top of 130 nm SiO₂ on a silicon wafer (hereafter: Wafer + 130).

Fig. 2(a) shows the efficiency of the aforementioned samples as a function of the silicon dioxide thickness measured at a wavelength of 780 nm. In agreement with finite difference time domain (FDTD) simulations, the data reveal a maximum efficiency around 75% for 105 nm SiO₂ (1st maximum) and 390 nm SiO₂ (2nd maximum) on a full-chip sized gold mirror as a bottom cavity. While similar results can be achieved for small gold mirrors with 105 nm SiO₂, the efficiencies for the data points in between are significantly

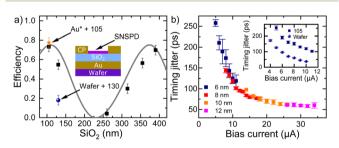


Fig. 2 Influence of multi-layer structure. (a) Measured and simulated efficiency of 8 nm NbTiN at 780 nm on gold mirror with varied SiO₂ thickness. Additional data points show the efficiency on a 100 μ m diameter gold mirror (hereafter: Au*) with 105 nm SiO₂ and on top of 130 nm SiO₂ on a silicon wafer (hereafter: Wafer + 130). The measured efficiencies agree very well with the simulated values. The inset depicts the multi-layer structure used including the (10 nm Ti/50 nm Au) contact pads. (b) Timing jitter depending on the applied current using different NbTiN thicknesses on gold mirrors with a 105 nm SiO₂ layer on top and measured at a wavelength of 850 nm. The jitter decreases with an increased current converging against 65 ps. The inset shows a comparison in the range from 4 μ A to 12 μ A of the same data with detectors directly fabricated on a Si/SiO₂ wafer. We observe that the jitter for the detectors on a Si/SiO₂ wafer are more than 50 ps lower compared to those on a full-chip sized (FCS) mirror with 105 nm SiO₂ on top.

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lower in agreement with FDTD simulations. The inset depicts a sketch of the multi-layer structures used including the (10 nm Ti/ 50 nm Au) contact pads. In Fig. 2(b) the timing jitter is shown as a function of the applied bias current for different NbTiN thicknesses on a FCS gold mirror with 105 nm SiO₂. The jitter throughout this work was measured at a wavelength of 850 nm and decreases with an increased current converging against 65 ps. The inset shows the same data in comparison to detectors fabricated on a Si/SiO₂ wafer in the range up to 12 μ A. Here, a significantly improved jitter can be observed for the detectors without Au mirror, reaching values of 37 ps. Therefore, the comparison indicates a strong influence of the gold mirror underneath.

3.1 Influence of the multi-layer structures

To analyze the influence of the multi-layer structures, detector pulses for 105 nm, 260 nm, 390 nm SiO_2 on a FCS gold mirror and 105 nm SiO_2 on the small gold mirror are shown in Fig. 3(a).

The pulses exhibit an increased steepness (from bottom to top) resulting not only in a shorter rise time but also in higher pulses. Note that the pulse for the 105 nm SiO₂ on a small gold mirror shows a significantly steeper rising edge compared to the same SiO₂ thickness on a FCS film. Moreover, for a FCS film a general trend of faster rise times for thicker SiO₂ layers can be observed. In Fig. 3(b) we present the corresponding pulse height as a function of the multi-layer structure used at a fixed bias current of 16 µA. From 0.39 V, the pulse height increases with an increasing SiO₂ thickness until it reaches the same pulse height of about 0.47 V as both the detector on the small gold mirror (inset Fig. 3(b)) and on Si/SiO₂ wafers. To conclude, for thicker SiO₂ thicknesses (or small gold mirrors) the rise time becomes shorter and the pulse height larger. To quantify these findings, we plot the rise time and timing jitter for the different samples at a fixed bias of 16 μ A in Fig. 4(a). Here, the timing jitter follows the same trend as the rise time. While for 105 nm SiO₂ on an FCS film a timing jitter of only

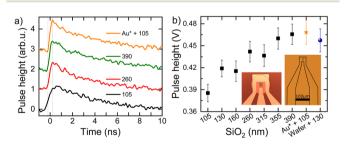


Fig. 3 Comparison of the detector response for different multi-layer structures at an applied bias current of 16 μ A. (a) Normalized voltage pulses shown for 105 nm, 260 nm, 390 nm SiO₂ on a FCS gold mirror and 105 nm SiO₂ on a small gold mirror. The reference lines mark the start and end position of the rising edge for the 100 μ m mirror. The rise time decreases and the pulse height increases with an increasing SiO₂ thickness. (b) Mean pulse height for detectors fabricated on different material combinations. The pulse height increases with an increasing SiO₂ thickness and for the detectors fabricated on wafer and small gold mirror. The inset depicts microscope images of the detector on both gold mirror types.

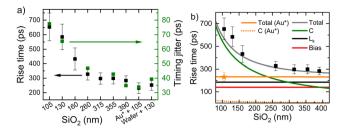


Fig. 4 Influence of distance and size of the buried gold mirror relative to the contact pad at a bias current of 16 μ A. (a) Rise time and timing jitter for the different multi-layer structures used. We observe a parallel decrease of both the rise time and jitter with an increased SiO₂ thickness or small gold mirrors. While the rise time decreases from 659 ps to 231 ps, the jitter improves from 77 ps to 33 ps. (b) Calculated rise time consisting of a capacitance (*C*), bias current (Bias) and kinetic inductance (*L*_k) component. While the kinetic inductance and bias component are the same for all material combinations, an additional capacitance contribution is formed between the contact pads and the FCS gold mirrors in agreement with the measured rise times.

80 ps was measured, we achieved around 30 ps for the same SiO₂ thickness on a small mirror. Again, the detectors on 390 nm SiO₂/Au, a small gold mirror and the detectors directly fabricated on the wafer behave similar confirming the results in Fig. 3. Fig. 4(b) depicts the rise time as a function of the SiO₂ thickness for different multi-layer structures. The bias (red) and kinetic inductance component (black) are constant at a fixed bias level (here: 16 μ A) and the same for both types of mirrors. The capacitance component on the other hand increases with an increasing SiO₂ thickness, but also strongly depends on the mirror design (dashed orange and solid green lines for Au* and full-chip size mirrors, respectively). The resulting total rise times for Au* mirror (solid orange line) and full-chip size mirror (solid gray line) can be calculated using $\tau_{tot} = \sqrt{\tau_{L_k}^2 + \tau_{Dias}^2 + \tau_{C}^2}$.³³

The fraction of the kinetic inductance $(\tau_{L_k} = \alpha \times L_k/R_n^{34})$ depends on the scaling factor $\alpha = 0.9-1/e$, which relates to the limited rising edge from 1/e to 90%, the kinetic inductance, and the resistance of the normal-conducting R_n region upon photon absorption. The normal conducting resistance can be calculated as $R_{\rm n} = \frac{V_{\rm HS}}{V_{\rm SNSPD} \times R_{\rm device}}$. With the hotspot volume $V_{\rm HS}$,³⁵ the SNSPD volume $V_{\rm SNSPD}$ depending on the device geometry and the measured resistance R_{device} of the presented detectors we determine it to be $R_n = 0.56 \text{ k}\Omega$. The second rise time component is related to the bias current. It decays with $\tau_{\rm bias} \propto \sqrt{1/I_{\rm Bias}}^{36}$ (cp. Fig. 5(a)) but stays constant at a fixed bias. To derive it, we considered detectors on a small mirror with a known kinetic inductance component. The third parameter is a capacitive component forming between the FCS (small) gold mirror and the gold contact pads. For each type of mirror we assume a plate capacitor $\frac{1}{C_{\text{tot}}} = \frac{1}{C_{\text{Sig}}} + \frac{1}{C_{\text{Gnd}}} = \frac{d_{\text{SiO}_2}}{\varepsilon_0 \varepsilon_{\text{R}}} \left(\frac{1}{A_{\text{Sig}}} + \frac{1}{A_{\text{Gnd}}}\right) \text{ considering both con-$

tributing areas signal (A_{Sig}) and ground (A_{Gnd}) separately, with

Fig. 5 Current dependence of different pulse parameters. (a) Current dependence of the pulse height (top), the timing jitter (middle) and the rise time (bottom) of the detector on a gold mirror with a diameter of 100 μ m. While the pulse height increases linearly, both the rise time and jitter decrease with an increasing bias current. (b) Ratio of rise time and timing jitter as a function of the applied current. Due to the large number of material combinations, the data set was divided between the bottom and top panel. The overall trend shows a decrease of the ratio from 0.3 at 5 μ A down to 0.1 above 20 μ A regardless of the material layer combination and only depending on the applied current.

 d_{SiO_2} as the thickness of the silicon dioxide layer, ε_0 as the vacuum permittivity and ε_{R} as the relative permittivity of silicon dioxide. Subsequently, we calculate the cutoff frequency $f_{\text{cutoff}} = \frac{1}{2\pi C_{\text{tot}} Z_{\text{load}}}$ and get the capacitive component of the rise

time $\tau_{\rm C} = \frac{1}{2f} \times \alpha$. The difference in $\tau_{\rm C}$ (orange dashed *vs.* green

line) shown in Fig. 4(b) is due to the different areas contributing to the capacitance. Thus, by choosing the small gold mirror, the rise time and therefore timing jitter can be improved significantly.

3.2 Current dependence of rise time, jitter and pulse height

After the impact of the multi-layer structure, we now consider the current dependence on the timing characteristics. To this end Fig. 5(a) shows a typical detectors dependence of the rise time, timing jitter and pulse height on the applied bias current. The data reveals a linear increase for the pulse height, an exponential decrease in the timing jitter (compare Fig. 2(b)) and a slow decrease of the rise time. Considering the latter two, we present the ratio of timing jitter and rise time as a function of the applied bias current in Fig. 5(b).

The data was divided between the top and bottom panel due to the large number of material combinations. Interestingly, all data points of the different multi-layer structures follow the same trend. This is very surprising as the rise time in Fig. 4(a) differs strongly for the different samples. However, it can be concluded that the ratio of timing jitter and rise time is stable for a fixed bias current (within a current dependent limit) regardless of the material combination used. This is an important finding for the characterization of detectors, as it allows to approximate the jitter for a given bias current and rise time.

3.3 Ratio of timing jitter and slew rate

To investigate the fundamental relation between pulse shape and timing jitter further, we look at the dependence of the timing jitter relative to the slew rate (ratio between pulse height and rise time) shown in Fig. 6(a).

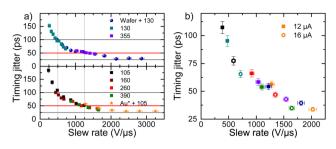


Fig. 6 Timing jitter as a function of the slew rate. (a) Due to the large number of material combinations, the data set was divided between the bottom and top panel. Both curves show the same trend of an improved jitter with an increasing slew rate, approaching a jitter of 25 ps. Reference lines at 500 V μ s⁻¹ and 1250 V μ s⁻¹ show a timing jitter of 100 ps and 50 ps, respectively, regardless of the material combination used. (b) Dependence of the timing jitter on the slew rate for different bias currents. The slew rate increases with an increased bias current at constant SiO₂ thickness. Furthermore, the slew rate also increases with increasing SiO₂ thickness at the same bias current. Hence, it is a combination of bias current and material combination that leads to higher slew rates and, consequently, to an improved timing jitter.

For an increased slew rate we observe a decreased timing jitter in agreement with D. Zhu.37 It decreases rapidly from 180 ps to 100 ps when the slew rate increases from 250 V μ s⁻¹ to 500 V µs⁻¹. Subsequently, the decrease slows down over 50 ps at 1250 V μ s⁻¹ until it reaches a timing jitter around 25 ps for a slew rate of 3000 V μ s⁻¹. Again, these findings are independent of the multi-layer structure used. Importantly, only the detectors fabricated on thick SiO₂ layers, wafer and small gold mirrors allow to achieve high slew rates and therefore low timing jitter values. The clear dependence of the timing jitter on the slew rate allows to deduce the timing jitter only by measuring the slew rate. Thus, by measuring a single electrical pulse (e.g. pulse height and rise time) it is possible to predict the timing jitter within a slew rate dependent tolerance using CW excitation, eliminating the need of a pulsed laser for jitter measurements if a rough estimate of the timing jitter is sufficient. Hence, such a scheme can help to speed up the characterization process for many detectors after a given setup has been calibrated. Fig. 6(b) shows a more detailed view of selected points at a bias current of 12 µA and 16 µA. For an increased bias current, the slew rate becomes higher and the timing jitter lower. Note that only the slew rate itself is strongly dependent on the multi-layer structure used. We conclude that a combination of bias current and material combination leads to higher slew rates and thus improved timing jitter. This fits well with the results in Fig. 3 and 4(a), where a decreased rise time was always accompanied by an increased pulse height resulting in a higher slew rate.

4 Conclusions

In summary, we investigated the relation between the SNSPD timing jitter and the detector response at a wavelength of

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850 nm measured at 4.5 K as well as the detection efficiency at 780 nm. The detectors were fabricated on a full-chip sized gold mirror (10 nm Ti/50 nm Au) with varying SiO₂ layer thicknesses, on small (100 µm diameter) gold mirrors with 105 nm SiO₂, and on Si/SiO₂ wafers. Besides a strong dependence on the bias current with an improved jitter for an increased current, we observed a dependence of the timing jitter on the multi-layer structure underneath the SNSPD. In particular, the entire pulse shape changes depending on the multi-layer structure used at a fixed bias current of 16 µA. This results in an increased pulse height with an increased SiO₂ thickness as well as for detectors on small gold mirrors or directly on a Si/SiO₂ wafer. Additionally, we investigate the rise time and timing jitter for the different material combinations at the same bias current. We observed that with an increased SiO₂ thickness and for small gold mirrors the rise time decreases leading to an improved timing jitter. Note that the origin of this is a capacitive behavior between the FCS gold mirror and the gold contact pads, which can be improved by using a thicker SiO₂ layer leading to a higher pulse height and faster rise time. However, by fabricating the detector on a small gold mirror (100 µm diameter) underneath the detector (and not the entire Ti/Au contact pad) it is possible to restore the timing jitter as found on bulk Si/SiO₂ substrates while maintaining the detection efficiency. Hence, to combine a good efficiency with a good timing jitter either a small gold mirror or a thick SiO₂ layer (>300 nm) has to be used. The second major finding is the relation between the timing jitter and the rise time. In particular, we were able to extend these findings to the dependence of the timing jitter on the slew rate describing the ratio between the pulse height and rise time. We found a general trend of the timing jitter to the slew rate, which is independent of the multi-layer structure used. We conclude that this enables us to estimate the timing jitter for a given detector pulse slew rate if a rough estimate of the timing jitter is sufficient. Thus, by only measuring single detector pulses this technique allows it to speed up the characterization process and paves the way for industrially scalable timing jitter measurements for instance with CW lasers.

Author contributions

R. F. designed study, analyzed data and drafted manuscript. R. F., L. Z., C. S. and S. S. fabricated samples. R. F. and C. S. performed the experiments. F. F. contributed simulations. J. F. and K. M. led the research projects. All authors discussed results and revised the manuscript.

Conflicts of interest

There are no conflicts to declare.

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